Monitoring of Dielectric Losses and Use of Ferrofluids for Bushing Cooling

S. D. Nedelcut, V. Proca, and D. M. Purcaru

Abstract-At present, the tendency to implement the conditionbased maintenance (CBM), which allows the optimization of the expenses for equipment monitoring, is more and more evident; also, the transformer substations with remote monitoring are increasingly used. This paper reviews all the advantages of the on-line monitoring and presents an equipment for on-line monitoring of bushings, which is the own contribution of specialists who are the authors of this paper. The paper presents a study of the temperature field, using the finite element method. For carrying out this study, the 3D modelling of the above mentioned bushing was performed. The analysis study is done taking into account the extreme thermal stresses, focusing at the level of the first cooling wing section of the ceramic insulator. This fact enables to justify the tand variation in time, depending on the transformer loading and the environmental conditions. With a view to reducing the variation of dielectric losses in bushing insulation, the use of ferrofuids instead of mineral oils is proposed.

Keywords-Monitoring, dielectric losses, ferrofluids, bushing.

I. INTRODUCTION

POWER transmission and distribution systems from the economically developed countries are already aged being, in many cases, more than 30 years old. All the fixed assets of the power systems have a standardized life time of maximum 30 years. From now on, any equipment is in danger to fail at any times with no previous warning. Under these conditions the on-line monitoring is one of the most useful methods for prolonging the life time up to the failure time limit and for replacing the assets or recovering them in the last moment. Because the bushing is a key element in the operation of electrical power equipment, its safe operation is a special issue because its degradation leads, in many cases, to the explosions followed by fires with very grave consequences for the stations and the substations. If up to the present the timebased maintenance (TBM), with periodical revisions, was used, in the last years one tries to find solutions to pass to the condition-based maintenance (CBM) which allows the minimization of the costs, the prolongation of equipment

Manuscript Received March 25, 2008.

V. Proca is with Research and Development and Testing National Institute for Electrical Engineering-ICMET Craiova (e-Mail: Vproca@Icmet.Ro).

D. M. Purcaru is with the University of Craiova, Romania (e-Mail: Dpurcaru@Yahoo.Com).

life time and the decrease of the risks of failure in exploitation with grave consequences. Based on the authors' experience in the field of monitoring of electric networks parameters and power transformers thermal condition [2], digital equipment for the monitoring of bushings insulation, with an original conception, was finalized.

II. CONTRIBUTIONS TO THE CONDITION-BASED MAINTENANCE OF THE BUSHINGS BY THEIR MONITORING

А. The Importance of On-Line Monitoring of Bushings

The damage of bushings is one of the main causes leading to the improper operation of the transformers or even to the explosions. The statistics confirmed that 30% of the transformer damages are due to capacitor-type bushings. The European statistics show that 80% of the damaged bushings are between 12 and 20 years old and therefore the monitoring is necessary even before the middle of their life time [1].

The high electric field gradients in the bushing insulation and the high working temperature contribute to the acceleration of insulation ageing.

The explosion of the bushing can damage the transformer tank, can generate an extended fire by transformer oil ignition, and fire in the bundle of cables in the electric switch box or in the control room through the secondary wiring.

A damage of the bushing leads to financial losses (between 1 and 3 million of dollars) to the insurers, both for physical damages and for the disturbances of the affairs in the companies they are serving.. These losses can reach, in exceptional situations, tens million of dollars.

The explosion can generate material damages and human life losses because of the porcelain pieces spread at long distance and with a very high speed.

The traditional diagnosis systems of the bushing insulation are based on periodical measurements of insulation loss factor, once within 2-3 years. In such case, it is necessary to put the transformers out of service and to measure tand at an applied voltage of 10 kV. The disadvantages of this traditional method for monitoring the bushing insulation are the following:

-The testing frequency arbitrarily chosen is not usually correlated with the failure rate development. The practice proved that the period between the measurements must not exceed 100 days to detect 95% from the defective bushings, and this is practically unacceptable;

S. D. Nedelcut is with the Research And Development And Testing National Institute For Electrical Engineering-ICMET Craiova, Romania (Corresponding Author : phone: +40 351 404 889; fax: +40 351 404 890; e-Mail: Dnedelcut@Icmet.Ro).

-The measurements for tan δ performed at an applied voltage of 10 kV are not relevant for the actual condition of the bushing insulation. The measurements at rated voltage, performed on the bushings where partial discharges appear, showed values of 5-8 times higher than those measured at 10 kV. The oil deterioration at high temperatures generates chemical modifications and sediment accumulations leading to the failures of the bushings. The detection of this type of fault at the voltage of 10 kV can be very difficult, even by tan δ measurement at the rated voltage.

-The traditional testing methods require a lot of work and the putting out of service for a long time.

By these reasons it is preferable to use on-line monitoring methods for the bushings.

Bushing monitoring methods known at present are presented below:

-use of a modified Schering bridge, where special methods are taken for suppressing the disturbances (standard capacitor supplied at the low voltage side of a voltage transformer connected on the same phase as the tested object, replacement of the conventional null indicator by a microcontroller-based measuring device fitted out with adaptive filtration). The equipment has in its structure standard components which require special metrological verifications, is bulky and could be achieved with high costs.

The equipment achieved by this method implementation allows monitoring the dielectric dissipation factor (tan δ), but is very expensive.

- vector addition of the capacitive currents of the bushings mounted on the three phases and detection of the unbalance current .



Fig. 1 Block diagram of capacitive current acquisition

Where:

A, B, C – bushings under potential

 C_1 - main capacity of the bushing

 C_2 - capacity of the last layer of the bushing

 $I_{\rm A}, I_{\rm B}, I_{\rm C}$ - capacitive currents acquired at the bushing measuring terminals

 I_{Σ} - vector adder

On the basis of the analysis on the magnitude and phase of the unbalance current ΔI_{Σ} the degree of modification for the dielectric properties of bushings is found; the disadvantage is that it does not present on line the value of tan δ for each bushing.



(b) determination of the initial unbalance current, $I^{0}{}_{\Sigma}$;

(c) determination of the unbalance current I_{Σ} , and sum current ΔI_{Σ} , taking also into account the initial unbalance current I^{0}_{Σ} ;

(d) determination of the unbalance current ΔI_{Σ} , modulus and phase, related the phase B

Fig. 2 Determination of the unbalance current ΔI_{Σ}

III. DIGITAL EQUIPMENT FOR ON-LINE MONITORING OF BUSHINGS

For on-line monitoring of bushings it is ideal to monitor the time variation of dielectric losses and of bushings own capacity.

In a quasi-homogeneous dielectric, in homogeneous electric field, the losses in dielectric depend on the electric field strength and temperature. The losses increase proportionally to the square of the electric field strength but they strongly depend also on the temperature θ in dielectric [6], [8].

$$P(\theta) = \frac{1}{2} \varepsilon_{0} \varepsilon_{r}(\theta) \cdot E^{2} \cdot \omega \tan \delta(\theta)$$
(1)

At high electric fields, the losses have even more accentuated increase related to the electric field strength.

The heat produced by Joule losses from the central conductor of the bushing generate also a temperature rise in the insulating material, which overlaps on that one due to the dielectric losses.

Because of dielectric losses, the alternating current which passes through a bushing is not purely reactive I_{r_i} but it has an active component I_a named leakage current.

Let us consider an elementary capacitor with the aria of the armature S, the thickness d, at the voltage U and the frequency f, which has the losses P:

$$\mathbf{P} = \mathbf{p}_{\theta} \cdot \mathbf{v} \cdot \mathbf{E}^{2} \tag{2}$$

where : p_{θ} are the specific volume losses (v = the volume of dielectric) at the temperature θ and, at the stress with the electric field having the strength E:

$$I_{a} = \frac{P}{U} = \frac{p_{\theta} \cdot \mathbf{v} \cdot \mathbf{E}^{2}}{\mathbf{E} \cdot \mathbf{d}} = \frac{p_{\theta} \cdot \mathbf{S} \cdot \mathbf{d} \cdot \mathbf{E}^{2}}{\mathbf{E} \cdot \mathbf{d}} = p_{\theta} \cdot \mathbf{S} \cdot \mathbf{E}$$
(3)

$$I_{r} = 2 \cdot \pi \cdot f \cdot U \cdot C = 2 \cdot \pi \cdot f \cdot E \cdot d \cdot \frac{S}{d} \cdot \varepsilon_{0} \cdot \varepsilon_{r} =$$

$$= 2 \cdot \pi \cdot f \cdot \varepsilon_{0} \cdot \varepsilon_{r} \cdot S \cdot E \qquad (4)$$

$$\tan \delta = \frac{I_a}{I_r} = \frac{p_\theta}{2 \cdot \pi \cdot f \cdot \varepsilon_0 \cdot \varepsilon_r}$$
(5)

The phasor diagram is:



Fig. 3 The phasor diagram

The dielectric losses of a bushing are:

 $\mathbf{P} = \mathbf{U} \cdot \mathbf{I}_{a} = \mathbf{U} \cdot \mathbf{I}_{r} \cdot \tan \delta = 2 \cdot \pi \cdot \mathbf{f} \cdot \mathbf{U}^{2} \cdot \mathbf{C} \cdot \tan \delta$ (6)

The used measuring diagram is the classical one:



 C_1, C_2 - bushing or CT under test R_1, R_2 - high voltage resistive divider (capacitive divider or voltage transformer)

r - additional shurt resistor

$$r \ll X_{C2}$$

After calibration $\phi[U_x, U_y] \sim k \text{ tg}\delta$

Fig. 4 The measuring diagram

For each monitored bushing the method presumes the acquisition of two signals: one signal taken over from the test tap of the bushing and the second signal, representing the reference voltage, taken over from the instrument transformer corresponding to the monitored bushing [3].

The taken over signals have non-sinusoidal periodical character and they had the following form:

$$f(t) = A_0 + \sum_{k=1}^{\infty} \left[M_k \cdot \cos(k\omega t) + N_k \cdot \sin(k\omega t) \right]$$
(7)

The implemented program performs the calculation of tan δ by the extraction of fundamentals from the sampled signals by a Fourier analysis algorithm.

The calculation algorithm presumes:

- determination of the coefficients M_k and $N_k \quad \mbox{for the fundamentals of the two signals } (k=1)$

$$M_{1} = \frac{2}{T} \int_{0}^{T} f(t) \cos(\omega t) dt$$
(8)

$$N_{1} = \frac{2}{T} \int_{0}^{T} f(t) \sin(\omega t) dt$$
(9)

The coefficients M_1 and N_1 are determined for each acquired quantity.

- determination of initial phases of the fundamentals

$$p = \arctan \frac{M_1}{M_1}$$
 (10) $q = \arctan \frac{M_1}{M_1}$ (11)

 $\varphi_1 = \arctan \frac{n_1}{N_1} \quad (10) \qquad \qquad \varphi_1 = \arctan \frac{n_1}{N_1} \quad (11)$

where,

 φ_1 = the initial phase of the fundamental of the signal taken over from the measuring terminal of the bushing;

 φ'_1 = the initial phase of the fundamental of reference signal taken over from the measuring terminals of the voltage transformer corresponding to the measured bushing.

Noting:
$$\delta_1 = \varphi_1 - \varphi_1'$$
 (12)

The loss dielectric factor is:

$$\tan \delta = \tan \left(90^\circ - \delta_1\right) \tag{13}$$

For non-sinusoidal state we have:

$$u = \frac{1}{C} \int i dt$$
 (14)

$$i = C \frac{du}{dt} = \sum_{i}^{\infty} \sqrt{2kC} \omega U_{k}$$
(15)

From here:

$$I_{k} = kC\omega U_{k}$$
(16)

Thus, by calculation, the own capacity of the bushing is obtained.



Monitorizare factor de pierderi dielectrice si curenti capacitivi

Fig. 5 Monitoring equipment mounted in the station

k

IV. CALCULATION OF BUSHING HEATING UNDER LONG TIME DUTY AND THERMAL CAPACITY

The calculation is performed in a covering manner, supposing that the axial thermal flux is neglected, but taking into account the Joule losses in the terminal and the dielectric losses in the capacitor, also their variation with temperature. The external temperature is considered as being given: θ_{e} .

Fig. 6 presents the calculation model. The rated current $I_{\rm N}$ flows through the terminal with diameter de, generating on the unit of length the Joule losses:

$$\mathbf{P}_{0} = \mathbf{k}_{\mathrm{r}} \cdot \boldsymbol{\rho}_{0} \cdot \left(\mathbf{l} + \boldsymbol{\alpha}_{0} \cdot \boldsymbol{\theta}_{0}\right) \cdot \frac{4}{\pi \cdot \mathbf{d}_{\mathrm{e}}^{2}} \cdot \mathbf{I}_{\mathrm{N}}^{2}$$
(17)



Fig. 6 Bushing calculation model

Here, k_r is the skin effect factor, ρ_0 is the resistivity of the terminal material and α_0 is the temperature coefficient at 0°C, and θ_0 is the terminal temperature in °C.

Further on, the increasing factor for the resistance of the solid circular cylindrical conductor is calculated.

One considers the case the solid circular cylindrical conductor with diameter d, resistivity ρ_r and temperature coefficient α_r at the reference temperature θ_r and relative permeability μ_r , in sinusoidal alternating current having the frequency f.

One calculates firstly the resistivity ρ_0 and the temperature coefficient α_0 at 0°C:

$$\rho_{0} = \rho_{r} \cdot (1 - \alpha_{r} \cdot \theta_{r}), \ \alpha_{0} = \frac{\alpha_{r}}{1 - \alpha_{r} \cdot \theta_{r}}$$
(18)

One notes:

$$\xi_0 = \frac{\mathrm{d}}{4} \cdot \sqrt{\frac{2 \cdot \pi \cdot \mathbf{f} \cdot \boldsymbol{\mu}_0 \cdot \boldsymbol{\mu}_r}{\rho_0}} \tag{19}$$

$$\xi = \frac{\xi_0}{\sqrt{1 + \alpha_0 \cdot \theta}} \tag{20}$$

if θ is the conductor temperature.

From the theory of the skin effect, the following expression of the complex inner impedance of the cylindrical conductor is set:

$$\underline{Z}_{i} = \mathbf{R}_{eli} + \mathbf{j} \cdot \mathbf{X}_{eli} = \mathbf{R}_{el0} \cdot \underline{\mathbf{k}}_{z}; \quad \underline{\mathbf{k}}_{z} = \frac{\zeta \cdot \mathbf{J}_{0} \cdot (2 \cdot \zeta)}{\mathbf{J}_{1} \cdot (2 \cdot \zeta)}$$
(21)

where $R_{el 0}$ is the direct current resistance of the conductor:

$$\mathbf{R}_{el_0} = \rho_0 \cdot \left(\mathbf{1} + \alpha_0 \cdot \theta\right) \cdot \frac{4 \cdot \mathbf{1}}{\pi \cdot \mathbf{d}^2}$$
(22)

By ζ the complex variable was symbolized:

$$\zeta = \xi \cdot j^{3/2} \tag{23}$$

and J_0 and J_1 are Bessel functions of first kind, order 0 and 1. Taking into account the shape of the series giving these functions:

$$J_{0} \cdot (2 \cdot \zeta) = \sum_{k=0}^{\infty} \frac{(-1)^{k}}{(k!)^{2}} \cdot \zeta^{2k}$$
(24)

$$J_{1} \cdot (2 \cdot \zeta) = \zeta \sum_{k=0}^{\infty} \frac{(-1)^{k}}{(k!)^{2} \cdot (k+1)} \cdot \zeta^{2k}$$

$$(25)$$

The following expression is derived:

$$\underline{\mathbf{x}}_{z} = \frac{1 - \zeta^{2} + \frac{1}{4} \cdot \zeta^{4} - \frac{1}{36} \cdot \zeta^{6} + \dots}{1 - \frac{1}{2} \cdot \zeta^{2} + \frac{1}{12} \cdot \zeta^{4} - \frac{1}{144} \cdot \zeta^{6} + \dots} = \frac{\sum_{i=0}^{\infty} \underline{\mathbf{a}}_{k}}{\sum_{i=0}^{k} \underline{\mathbf{b}}_{k}} = \frac{\sum_{i=0}^{\infty} (\mathbf{u}_{k} + \mathbf{j} \cdot \mathbf{v}_{k})}{\sum_{i=0}^{\infty} (\mathbf{x}_{k} + \mathbf{j} \cdot \mathbf{y}_{k})}$$
(26)

The coefficients of the series from the numerator and denominator are determined step by step, using the algorithm:

$$\underline{\mathbf{a}}_{1} = 1$$
 and then $\underline{\mathbf{a}}_{k+1} = -\underline{\mathbf{a}}_{k} \cdot \left(\frac{\zeta}{k+1}\right)^{2}$ si $\underline{b}_{k} = \frac{\underline{a}_{k}}{k+1}$ (27)

respectively $u_1 = 1$ and then

$$\mathbf{x}_{k} = \frac{\mathbf{u}_{k}}{2k+1}, \ \mathbf{v}_{k} = \xi^{2} \cdot \frac{\mathbf{x}_{k}}{2k+1}, \ \mathbf{y}_{k} = \frac{\mathbf{v}_{k}}{2k+2}, \ \mathbf{u}_{k+1} = -\xi^{2} \cdot \frac{\mathbf{y}_{k}}{2k+2}$$
(28)

⁽²⁸⁾ The series being with oscillating sign terms, they are added until the modulus of the added term decreases below the modulus of the sum multiplied with relative tolerance.

The algorithm is a stable one, in simple accuracy, up to $\xi = 5$, limit sufficient for practice. Further on, asymptotic developments could be used.

The factor for increasing the resistance, k_r, represents the real part of the factor k,

$$k_{r} = \frac{R_{eli}}{R_{el0}} = \text{Re} \cdot \left(\underline{k}_{z}\right)$$
(29)

Note – the factor \underline{k}_z has the significance $k_r + \frac{J \cdot X_0 \cdot K_x}{R_{el0}}$,

where x_0 is the inner reractance at direct current distribution:

 $\frac{x_0}{R_{el\,0}} = \frac{\xi^2}{2}$ to the cylindrical conductor. k_x is the factor for

changing the inner reactance of the cylindrical conductor, due to the skin effect.

Observation - For bushings with rated currents up to 2500 A the skin effect within the current path could be neglected.

The terminal is inside the guiding tube with inner diameter d_i. It results a thermal resistance per unit of length:

$$\mathbf{R}_{0} = \frac{1}{2 \cdot \pi \cdot \lambda_{u}} \cdot \ln \frac{\mathbf{d}_{i}}{\mathbf{d}e} \approx \frac{1}{\pi \cdot \lambda_{u}} \cdot \frac{\mathbf{d}_{i} - \mathbf{d}_{e}}{\mathbf{d}_{i} + \mathbf{d}_{e}}$$
(30)

Towards the environment, the terminal is bounded by a flange with outer diameter d_f and it transmits heat to the environment with temperature θ_e by a release factor α_e (16 -32 W/m² C). It results a thermal resistance per unit of length

International Journal of Electrical, Electronic and Communication Sciences ISSN: 2517-9438 Vol:2, No:4, 2008

$$R_{e} = \frac{1}{\pi \cdot \alpha_{e} \cdot d_{f}}$$
(31)

If the bushing flange is fixed on the transformer tank, one could consider $R_e \approx 0$, and for θ_e the maximum temperature of the oil at the level of transformer cap ($\theta_e \approx 90^{\circ}$ C) is taken.

The capacitor bushing itself is made of n layers numbered 1....n, from inside towards outside. The layer numbered by k is comprised between the diameters d_{k-1} and d_k , has a middle temperature θ_k , the voltage U_k is applied on it; dielectric losses per unit of length, P_k , are developed within the layer. On each diameter d_k an enclosure (foil) with axial length l_k is located. The geometric permeance per unit of length for the layer k is symbolized by Λ_k :

$$A_{k} = \frac{2\pi}{\ln \frac{d_{k}}{d_{k-1}}} \approx \pi \cdot \frac{d_{k} + d_{k-1}}{d_{k} - d_{k-1}}, \ k \in (1, n)$$
(32)

The dielectric losses per unit of length for the layer k are expressed as:

$$\mathbf{P}_{k} = \mathbf{p}_{d} \cdot \boldsymbol{\Lambda}_{k} \cdot \mathbf{e}^{\sigma(\theta_{k} - \theta_{d})} \cdot \mathbf{U}_{k}^{2}, \ k \in (1, n)$$
(33)

where p_d is the volume density of the dielectric losses at reference temperature θ_d and for the unit of electric field strength, and σ is the temperature coefficient for the exponential variation of specific losses.

Note - In general, the relation from below is valid:

$$\mathbf{p}_{d} = 2 \cdot \pi \cdot \mathbf{f} \cdot \boldsymbol{\varepsilon}_{0} \cdot \boldsymbol{\varepsilon}_{r} \cdot \mathbf{tg}\delta \tag{34}$$

where f is the frequency, ε_0 is the vacuum permittivity, ε_r is the dielectric relative permittivity and δ is the loss angle.

The voltage U_k applied to the layer results from the voltage U_m applied to the bushing.

$$U_{k} = U_{m} \cdot \frac{S_{k}}{\sum_{i=1}^{n} S_{i}}, \ k \in (1, n)$$

$$(35)$$

where by S_k the elastance of the layer k was expressed:

$$\mathbf{S}_{k} = \frac{1}{\boldsymbol{\varepsilon}_{0} \cdot \boldsymbol{\varepsilon}_{k} \cdot \boldsymbol{\Lambda}_{k} \cdot \mathbf{l}_{k}}, \quad \mathbf{k} \in (1, n)$$
(36)

By ε_k it was noted the relative permittivity of the dielectric from layer k, which may depend on the temperature by means of the relation:

$$\varepsilon_{k} = \varepsilon_{d} \cdot e^{\zeta(\theta_{k} - \theta_{d})}, \ k \in (1, n)$$
 (37)

 ε_{d} is the relative permittivity of the dielectric at the reference temperature θ_{d} , and ζ is the temperature coefficient.

The thermal resistance of the layer k is expressed by the relation:

$$R_{k} = \frac{1}{\lambda \cdot \Lambda_{k}}, \ k \in (0, n+1)$$
(38)

where λ is the thermal conductibility of the dielectric.

Fig. 7 presents the diagram for bushing heating calculation: One makes the notations:

$$Z_{0} = R_{0} + R_{0} + \frac{R_{1}}{2}; \ Z_{k} = \frac{R_{k} + R_{k+1}}{2}, \ k \in (1, n-1); \ Z_{n} = \frac{R_{n}}{2} + R_{e} + R_{e}^{'}$$
(39)



Fig. 7 Diagram for bushing heating calculation

where by R_0 si R_e there were expressed the thermal resistances of the possible buffer layers, no voltage, from bushing inside and outside (39).

Temperatures are immediately determined by the relation:

$$\theta_{k} = \theta_{k+1} + Z_{k} \cdot \sum_{i=0}^{k} P_{i}, \ k \in (0, n)$$

$$(40)$$

Because losses depend on temperature, directly or by means of voltage distribution on layers, the resolution (40) is iteratively resumed, correcting each time the losses in accordance with the previously calculated temperatures, until a stationary solution (fixed point) is got.

V. DEPENDENCE OF DIELECTRIC LOSS VARIATION ON THE LOAD FACTORS AND ENVIRONMENTAL CONDITIONS

By analysing the got results, it is noticed the identical behaviour of the dielectric layers, namely a significant increase of the average temperature, due to the increase of the environmental temperature, especially in the hours of load peak (17 - 22), even when the environmental temperature decreases significantly. But the material parameter depends also on the temperature, according to the relation:

$$\overline{\mathbf{p}}_{\theta} = \overline{\mathbf{p}}_{\theta_0} \cdot \exp[\mathbf{a}(\theta - \theta_0)] \tag{41}$$

where "a" is a constant depending on the dielectric nature.

Result:
$$\tan \partial = \tan \partial_0 \exp[a(\theta - \theta_0)]$$
 (42)

For the bushings with resin-lacquered paper insulation , as a results of the laboratory experiments, the variation of $tan\delta$ as a function of temperature is:



Fig. 8 The variation of tan δ as a function of temperature

From here, the explanation of the very high variation of the dielectric losses in the course of one day and, especially, the dependence on the network load variation can be very clearly obtained.

The application of this information is very useful because it confirms this accuracy of the measurements performed with this monitoring equipment placing confidence in this equipment utilization for protecting the capacitor-type bushings, the high power transformer, key-factors in the good operation of power systems, implicitly.

VI. USE OF FEROFLUIDS FOR BUSHING COOLING

As it results from the previous chapter, dielectric losses are closely depending on the temperature of the insulating body. The authors proposed themselves to study the improvement of the bushing cooling, by using ferrofluids instead of mineral oils.

Ferrofluids are suspensions of single domain magnetite particles with average diameters of approximately 10 nm stabilized by surfactancts in carrier liquids. Ferrofluids have found application in a variety of engineered devices and systems for, among other things, lubrication and sealing of bearings. Recently, their potential utility in cooling of electrical equipments has been recognized.

It is known the change of the flowing depending on temperature, under the presence of electromagnetic field, this fact leading obviously to the improvement of cooling.

Taking into account the thermal model presented in Chapter 4, the overtemperature distribution on the bushing radius is got.

Laboratory experiments have been performed and an improvement of the cooling efficiency was found, decreasing the temperature of the insulating body by up to 10%, this fact leading to a much lower variation of tan δ with temperature.

REFERENCES

- Arshad M., Islam S.M., Power transformer condition monitoring and assessment for strategic benefits, Curtin University of Technology Department of Electrical & Computer Engineering, Perth, Australia.
- [2] Nedelcut D., Popa D., Monitoring, Control and Protection Equipment for Power Transformers, Workshop "Technic Qualification and Transient Behaviour of Components of High Voltage Substation", Karlsruhe-Germania, 2002.
- [3] Nedelcut D., Marinescu A., Sacerdoțianu D., Pătru I., Purcaru I., Equipment for HV Capacitor-Type Bushings (DPD) monitoring, "The 7th International Conference on Live Maintenance – ICOLIM – 2004", Bucureşti, România, 2004.
- [4] Nedelcut D., Bică M., Țălu M., Monitoring of Capacitor-Type Bushings Considerations Regarding the Influence on Thermal Field Distribution on Dielectric Losses, IASTED Conference – Power and Energy Systems EuroPES 2005, Benalmadena, Spain.
- [5] Nedelcut D., Marinescu A., Sacerdoțianu D., Pătru I., Purcaru I. -Equipment for on-line monitoring of capacitor-type bushings, CIGRE Working Group D1.02 Meeting, Romania, Eforie,2005.
- [6] Roth, A., Hochspannungstechnik, IV edition, Wien, 1959.
- [7] Wang N., Lu F.C., Li H.M., Analytical Processing of On-Line Monitored Dissipation Factor Based on Morphological Filter, IEEE, 2004.
- [8] Ifrim A., Notinger P., Electrotechnical Materials, EDP, Bucureşti, 1979.
 [9] Kuchler A., Neumann C., Koch N., Loppach K., Krause C., Alff J.J., Condition Assessment of Aged Transformer Bushing Insulations, CIGRE 2006.

- [10] Bica, M., Thermotechnics and heat engines, Universitaria Publishing House, Craiova, 2000.
- [11] Zalewcki, J., Izolacja przekladnikov pradowych na napiecia 220 i 400 kV, PET (1962), no 7, pages 273-281.
- [12] *** Cosmos M.2.6. User guide, Structural Research and Analysis Corporation, Los Angeles, California, 2000.
- [13] *** SolidWorks. User guide, Structural Research and Analysis Corporation, Los Angeles, California, 2004.
- [14] Arnim Nethe, Thomas Scholz, Hanns-Dietrich Stahlmann, Ferrofluid driven electric motors – heat budget and experimental verification, Workshop for ferrofluids, Benediktbeuern, 2000.
- [15] M.Timko, P.Kopcansky, K.Marton, L.Tomco, Dielectric properties of magnetic fluid, Institute of Experimental Physics Slovak Academy of Sciences- Kosice, Faculty of Electrical Engineering and Informatics – Kosice, Faculty of Aviation – Kosice.



Silvius-Dorel S. Nedelcut (born in Craiova, Romania, on October 22, 1953) received the Dipl.-Eng. and Ph. D. degrees in electrical engineering from the University of Craiova, Romania, in 1977 and 2007, respectively. He has a rich experience as research engineer in electric fields and he was published 3 books and more than 40 papers in technical journals and conference proceedings.

Her research interests include equipments for monitoring, testing, fault diagnosis and disturbance recording.

Dr. Nedelcut is a member of the TC 36 of IEC since 1992, a member of the "American Association for the Advancement of Science" since 2003 and is a reviewer for Journal of Materials Processing Technology since 2007.