Hybrid Algorithm for Hammerstein System Identification Using Genetic Algorithm and Particle Swarm Optimization

Tomohiro Hachino, Kenji Shimoda, and Hitoshi Takata

Abstract—This paper presents a method of model selection and identification of Hammerstein systems by hybridization of the genetic algorithm (GA) and particle swarm optimization (PSO). An unknown nonlinear static part to be estimated is approximately represented by an automatic choosing function (ACF) model. The weighting parameters of the ACF and the system parameters of the linear dynamic part are estimated by the linear least-squares method. On the other hand, the adjusting parameters of the ACF model structure are properly selected by the hybrid algorithm of the GA and PSO, where the Akaike information criterion is utilized as the evaluation value function. Simulation results are shown to demonstrate the effectiveness of the proposed hybrid algorithm.

Keywords—Hammerstein system, identification, automatic choosing function model, genetic algorithm, particle swarm optimization.

I. Introduction

ITH the advance of technology, the importance of non-linear modeling in control engineering has been growing. Since most practical systems have inherently nonlinear characteristics such as saturation and dead-zone, the development of accurate nonlinear system identification algorithm is a key problem for precise analysis, prediction or control design. One of approaches for nonlinear system identification is use of the block-oriented models. The Hammerstein model is one of the block-oriented models and has been utilized to express nonlinear systems in many applications such as predistorters in wireless communication systems [1] and dc/dc converters [2]. The model consists of a nonlinear static part followed by a linear dynamic part. It has many advantages for control design or stability analysis owing to the simple model structure [3]. Several identification algorithms have been investigated for the Hammerstein model by using correlation theory [4], neural networks [5], orthogonal functions [6], polynomials [7], piecewise linear model [8], automatic choosing function (ACF) model [9], and so on.

In this paper a hybrid algorithm for model selection and identification of discrete-time Hammerstein systems is proposed by using the genetic algorithm (GA) [10] and particle

Tomohiro Hachino is with the Department of Electrical and Electronics Engineering, Kagoshima University, Kagoshima, 890-0065 Japan (phone and fax: +81-99-285-8392; e-mail: hachino@eee.kagoshima-u.ac.jp).

Kenji Shimoda is with the Department of Electrical and Electronics Engineering, Kagoshima University, Kagoshima, 890-0065 Japan.

Hitoshi Takata is with the Department of Electrical and Electronics Engineering, Kagoshima University, Kagoshima, 890-0065 Japan (e-mail: takata@eee.kagoshima-u.ac.jp).

Manuscript received February 28, 2009.

swarm optimization (PAO) [11]. The unknown nonlinear static part to be estimated is represented by an ACF model. The weighting parameters of the ACF and the system parameters of the linear dynamic part are estimated by the linear leastsquares (LS) method. The adjusting parameters of the ACF model structure, i.e., the number and widths of the subdomains and the shape of the ACF, are properly determined by the hybrid algorithm of the GA and PSO. The GA is a probabilistic search procedure based on the mechanics of natural selection and natural genetics [10]. It is well known that although the GA has a high potential for global optimization, its convergence is quite slow. On the other hand, PSO is a swarm intelligence optimization technique which was inspired by the social behavior of a flock of birds or a shoal of fish, and has been empirically shown to be very efficient for optimization [11]-[14]. Some theoretical researches such as stability analysis and parameter selection have been also reported for PSO [15], [16]. Similarly to the GA, PSO searches from not a single point but a population of points and only uses information about the objective function, not derivatives or other auxiliary knowledge. PSO is simpler than the GA, because the algorithm of PSO consists of only the basic arithmetic operations and does not require complicated coding and genetic operations such as crossovers and mutations. PSO has quite fast convergence property, but it is inferior to the GA in global optimization. Therefore, we combine the GA with PSO to optimize the adjusting parameters of the ACF model structure. At first the GA are utilized to serve for global optimization, then PSO is applied to work for local optimization, where the Akaike information criterion (AIC) [17] is used for the objective function (fitness value function).

The outline of this paper is as follows: In section II the problem is formulated. In section III the identification method is proposed in the case of fixed ACF model structure. In section IV the optimization method for the adjusting parameters of the ACF model structure is considered by hybridization of the GA and PSO. In section V simulation results are shown to examine the effectiveness of the proposed hybrid algorithm. Finally conclusions are given in section VI.

II. STATEMENT OF THE PROBLEM

Consider a discrete-time nonlinear system described by the Hammerstein model shown in Fig. 1:

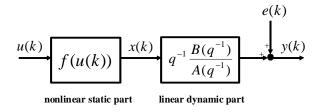


Fig. 1. Hammerstein model

$$\begin{cases}
A(q^{-1})y(k) = B(q^{-1})x(k-1) + A(q^{-1})e(k) \\
x(k) = f(u(k)) \\
A(q^{-1}) = 1 + a_1q^{-1} + \dots + a_nq^{-n} \\
B(q^{-1}) = b_0 + b_1q^{-1} + \dots + b_rq^{-r}
\end{cases}$$
(1)

where u(k) and y(k) are input and output signals, respectively. x(k) is an intermediate signal that cannot be measured. e(k)is the measurement noise. q^{-1} denotes the backward shift operator. n and r are the known degrees of polynomials $A(q^{-1})$ and $B(q^{-1})$, respectively. $f(\cdot)$ is an unknown nonlinear function. The problem is to identify the system parameters $\{a_i\}$ and $\{b_i\}$ of the linear dynamic part and the nonlinear static function $f(\cdot)$ from input and output measurements.

III. IDENTIFICATION

In this section the identification algorithm in the case of fixed ACF model structure is presented. In the next section the adjusting parameters of the ACF model structure will be determined by the hybrid approach.

In order to represent the nonlinear function $f(\cdot)$, the sigmoid type ACF [18] is introduced. Let a domain being a data region of u(k) be $D=[u_{min},u_{max}]$. The domain D is divided into some subdomains of $D=\cup_{i=1}^M D_i$ where $D_i=[\alpha_i,\beta_i], \ \alpha_1=[\alpha_i,\beta_i]$ u_{min} , $\beta_M = u_{max}$, $\alpha_k = \beta_{k-1}$ $(k = 2, 3, \dots, M)$. Then the ACF is defined by

$$I_{i}(u(k)) = 1 - \frac{1}{1 + \exp(H(u(k) - \alpha_{i}))} - \frac{1}{1 + \exp(-H(u(k) - \beta_{i}))}$$
(2)

where H is positive real value. $I_i(u(k))$ is almost unity only on a subdomain $D_i = [\alpha_i, \beta_i]$ and nearly equals to zero on $D-D_i$, so it chooses D_i automatically. The ACFs are shown in Fig. 2 in the case of H = 6,20 and 200.

Assume that f(u(k)) is well approximated linearly on each subdomain D_i :

$$f(u(k)) \simeq c_i + d_i u(k)$$
 on D_i . (3)

Then f(u(k)) is represented by using the ACF on the whole domain D as

$$f(u(k)) = \sum_{i=1}^{M} (c_i + d_i u(k)) I_i(u(k)) + \epsilon(k)$$
on D ,
$$(4)$$

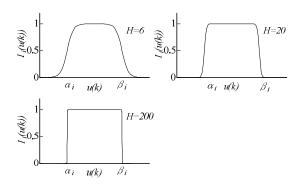


Fig. 2. Automatic choosing function (ACF)

where $\epsilon(k)$ is an approximation error. The schematic diagram of the ACF model is depicted in Fig. 3.

Substituting (4) into (1) yields

$$A(q^{-1})y(k) = \sum_{i=1}^{M} c_i B(q^{-1}) I_i(u(k-1))$$

$$+ \sum_{i=1}^{M} d_i B(q^{-1}) u(k-1) I_i(u(k-1)) + v(k)$$
(5)

or in vector form,

$$y(k) = \boldsymbol{\varphi}^{\mathrm{T}}(k)\boldsymbol{\theta} + v(k) \tag{6}$$

where $v(k) = A(q^{-1})e(k) + B(q^{-1})\epsilon(k-1)$ is the equation

$$\begin{cases} \boldsymbol{\theta} = [\boldsymbol{\theta}_{a}^{\mathrm{T}}, \boldsymbol{\theta}_{c_{1}}^{\mathrm{T}}, \boldsymbol{\theta}_{c_{2}}^{\mathrm{T}}, \cdots, \boldsymbol{\theta}_{c_{M}}^{\mathrm{T}}, \boldsymbol{\theta}_{d_{1}}^{\mathrm{T}}, \boldsymbol{\theta}_{d_{2}}^{\mathrm{T}}, \cdots, \boldsymbol{\theta}_{d_{M}}^{\mathrm{T}}]^{\mathrm{T}} \\ \boldsymbol{\theta}_{a} = [a_{1}, a_{2}, \cdots, a_{n}]^{\mathrm{T}} \\ \boldsymbol{\theta}_{c_{i}} = [\theta_{c_{i}}(1), \theta_{c_{i}}(2), \cdots, \theta_{c_{i}}(r+1)]^{\mathrm{T}} \\ = [b_{0}c_{i}, b_{1}c_{i}, \cdots, b_{r}c_{i}]^{\mathrm{T}} \\ \boldsymbol{\theta}_{d_{i}} = [\theta_{d_{i}}(1), \theta_{d_{i}}(2), \cdots, \theta_{d_{i}}(r+1)]^{\mathrm{T}} \\ = [b_{0}d_{i}, b_{1}d_{i}, \cdots, b_{r}d_{i}]^{\mathrm{T}} \\ \boldsymbol{\varphi}(k) = [\boldsymbol{\varphi}_{a}^{\mathrm{T}}(k), \boldsymbol{\varphi}_{c_{1}}^{\mathrm{T}}(k), \boldsymbol{\varphi}_{c_{2}}^{\mathrm{T}}(k), \cdots, \boldsymbol{\varphi}_{c_{M}}^{\mathrm{T}}(k), \\ \boldsymbol{\varphi}_{d_{1}}^{\mathrm{T}}(k), \cdots, \boldsymbol{\varphi}_{d_{M}}^{\mathrm{T}}(k)]^{\mathrm{T}} \\ \boldsymbol{\varphi}_{a}(k) = [-y(k-1), -y(k-2), \cdots, -y(k-n)]^{\mathrm{T}} \\ \boldsymbol{\varphi}_{c_{i}}(k) = [I_{i}(u(k-1)), I_{i}(u(k-2)), \\ \cdots, I_{i}(u(k-r-1))]^{\mathrm{T}} \\ \boldsymbol{\varphi}_{d_{i}}(k) = [u(k-1)I_{i}(u(k-1)), u(k-2)I_{i}(u(k-2)), \\ \cdots, u(k-r-1)I_{i}(u(k-r-1))]^{\mathrm{T}} \\ (i=1, 2, \cdots, M). \end{cases}$$

Thus, the unknown parameter vector θ is easily evaluated by applying the linear LS method to (6):

$$\hat{\boldsymbol{\theta}} = \left[\sum_{k=N_s+1}^{N_s+N} \boldsymbol{\varphi}(k) \boldsymbol{\varphi}^T(k) \right]^{-1} \left[\sum_{k=N_s+1}^{N_s+N} \boldsymbol{\varphi}(k) y(k) \right]$$
(8)

where N is the number of input and output data.

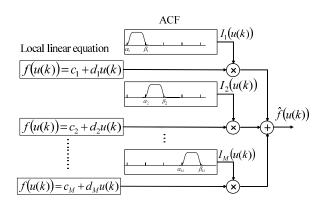


Fig. 3. ACF model

The parameters of the linear dynamic part are estimated by

$$[\hat{a}_1, \cdots, \hat{a}_n, \hat{b}_0, \cdots, \hat{b}_r]^{\mathrm{T}}$$

$$= \begin{bmatrix} \mathbf{I}_{(n+r+1)\times(n+r+1)} : & \mathbf{0} \end{bmatrix} \hat{\boldsymbol{\theta}},$$
(9)

setting $\hat{c}_1 = 1$ without loss of generality. Namely the parameters of the linear dynamic part can be obtained by taking the first n+r+1 elements of $\hat{\boldsymbol{\theta}}$.

Next, the parameters for the nonlinear static part are obtained by using the linear LS technique again as

$$\hat{c}_{i} = \sum_{j=1}^{r+1} \hat{\theta}_{c1}(j)\hat{\theta}_{ci}(j) / \sum_{j=1}^{r+1} \hat{\theta}_{c1}^{2}(j)$$

$$(i = 2, 3, \cdots, M)$$

$$\hat{d}_{i} = \sum_{j=1}^{r+1} \hat{\theta}_{c1}(j)\hat{\theta}_{di}(j) / \sum_{j=1}^{r+1} \hat{\theta}_{c1}^{2}(j)$$

$$(i = 1, 2, \cdots, M).$$

$$(10)$$

Thus the nonlinear static function is composed of the \hat{c}_i and d_i in (10) as

$$\hat{f}(u(k)) = \sum_{i=1}^{M} (\hat{c}_i + \hat{d}_i u(k)) I_i(u(k)).$$
 (11)

IV. HYBRID ALGORITHM OF GA AND PSO FOR MODEL SELECTION

The accuracy of the above identification algorithm greatly depends on the adjusting parameters of the ACF model structure, i.e., the number M, the widths $\{\alpha_i\}$ of the subdomains and the parameter H of the ACF. Therefore $X = [M, \{\alpha_i\}, H]$ should be properly determined. In this section the hybrid algorithm of the GA and PSO is proposed to determine X. The proposed hybrid algorithm has two stages. At the first stage, the GA is utilized until the prespecified generation to optimize X. Then at the second stage, PSO is applied to determine X, where the final population of the model candidates obtained by the GA is used as the initial population of particle positions.

The proposed algorithm is described as follows:

 $\frac{\textit{Step 1-1: Initialization for GA}}{\textit{Generate an initial population of } Q \textit{ individuals for } X}$ randomly.

Set the generation number g to 0.

Step 1-2: Decoding

Decode Q strings into real values X_i^g $(i = 1, 2, \dots, Q)$.

Step 1-3: Judgement of switch from GA to PSO

If the generation number q is greater than the prespecified q_{max} , then go to Step 2-1 of the stage 2 (PSO).

Step 1-4: Construction of ACF model

Construct Q candidates of the ACF model using X_i^g (i = $1, 2, \cdots, Q$).

Step 1-5: Identification by the LS method

Identify $\widehat{\theta}_i$ and $\widehat{f}_i(u(k))$ $(i = 1, 2, \dots, Q)$ from (8)~(11), using each candidate of the ACF model.

Step 1-6: Fitness value calculation

Calculate the evaluation values (AIC):

$$J_{i}(\boldsymbol{X}_{i}^{g}) = N \log \left\{ \frac{1}{N} \sum_{k=N_{s}+1}^{N_{s}+N} (y(k) - \hat{y}_{i}(k))^{2} \right\} + 2P_{i}$$

$$(i = 1, 2, \dots, Q)$$
(12)

and the fitness values $F_i(\mathbf{X}_i^g) = \exp(-J_i(\mathbf{X}_i^g)/300)$ using X_i^g . $P_i = n + M_i(r+1)$ is the number of the parameters in the identification model (6). $\hat{y}_i(k)$ is the output of the *i*th candidate of the estimated model.

Step 1-7: Reproduction

Reproduce each of individual strings with the probability of $F_i / \sum_{j=1}^Q F_j$. Practically, the linear fitness scaling [10] is utilized to avoid undesirable premature convergence.

Step 1-8: Crossover

Pick up two strings randomly and decide whether or not to cross them over according to the crossover probability P_c . Exchange strings at a crossing position if the crossover is required. The crossing position is chosen randomly.

Step 1-9: Mutation

Alter a bit of string (0 or 1) according to the mutation probability P_m .

Step 1-10: Repetition for GA

Set the generation number to g = g + 1 and go to Step 1-2.

Step 2-1: Initialization for PSO

Set an initial population of Q particles with positions X_i^0 and velocities V_i^0 $(i=1,2,\cdots,Q)$. The positions X_i^0 are constructed by the final population of Q individuals in the GA. The velocities V_i^0 are set to be random values.

Set the iteration counter l to 0.

Step 2-2: Construction of ACF model

Construct Q candidates of the ACF model using X_i^l (i = $1, 2, \cdots, Q$).

Step 2-3: Identification by the LS method

Identify $\hat{\theta}_i$ and $\hat{f}_i(u(k))$ $(i = 1, 2, \dots, Q)$ from (8)~(11), using each candidate of the ACF model.

Step 2-4: Evaluation value calculation

Calculate the evaluation values (AIC):

ISSN: 2517-9438

Vol:3, No:5, 2009

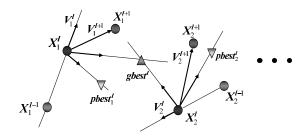


Fig. 4. Update of particles

$$J_{i}(\boldsymbol{X}_{i}^{l}) = N \log \left\{ \frac{1}{N} \sum_{k=N_{s}+1}^{N_{s}+N} (y(k) - \widehat{y}_{i}(k))^{2} \right\} + 2P_{i}$$

$$(i = 1, 2, \dots, Q).$$
(13)

Step 2-5: Update of the best positions pbest and gbest

Update $pbest_i^l$, which is the personal best position, and $gbest^l$, which is the global best position among all particles as follows:

If l = 0 then

$$egin{aligned} egin{aligned} egin{aligned\\ egin{aligned} egi$$

otherwise

$$\begin{aligned} \boldsymbol{pbest}_{i}^{l} &= \left\{ \begin{array}{ll} \boldsymbol{X}_{i}^{l} & (\ J(\boldsymbol{X}_{i}^{l}) < J(\boldsymbol{pbest}_{i}^{l-1}) \) \\ \boldsymbol{pbest}_{i}^{l-1} & (\ \text{otherwise} \) \\ \boldsymbol{gbest}^{l} &= \boldsymbol{pbest}_{i_{best}}^{l} & i_{best} = arg \min_{i} J(\boldsymbol{pbest}_{i}^{l}). \end{array} \right. \end{aligned}$$

Step 2-6: Update of positions and velocities

Update the particle positions and velocities using (16):

$$\left\{ \begin{array}{l} \boldsymbol{V}_{i}^{l+1} = \boldsymbol{w} \cdot \boldsymbol{V}_{i}^{l} + \boldsymbol{c}_{1} \cdot rand_{1}() \cdot (\boldsymbol{pbest}_{i}^{l} - \boldsymbol{X}_{i}^{l}) \\ + \boldsymbol{c}_{2} \cdot rand_{2}() \cdot (\boldsymbol{gbest}^{l} - \boldsymbol{X}_{i}^{l}) \\ \boldsymbol{X}_{i}^{l+1} = \boldsymbol{X}_{i}^{l} + \boldsymbol{V}_{i}^{l+1} \end{array} \right.$$

$$(16)$$

where w is an inertia factor, c_1 and c_2 are constants representing acceleration coefficients, and $rand_1()$ and $rand_2()$ are uniformly distributed random numbers with amplitude in the range [0,1]. Figure 4 shows the update of particle positions.

Step 2-7: Repetition for PSO

Set the iteration counter to l = l + 1 and go to Step 2-2 until the termination criterion is satisfied.

Finally, at the termination of this algorithm when $l = l_{max}$, the suboptimal adjusting parameter vector \hat{X} of the ACF model structure is determined by the best position $gbest^{l_{max}}$. Thus, the final estimated model is constructed from \hat{X} and the corresponding $\hat{\theta}$ and $\hat{f}(u(k))$.

For clarity, the flow chart of the proposed hybrid method is shown in Fig. 5.

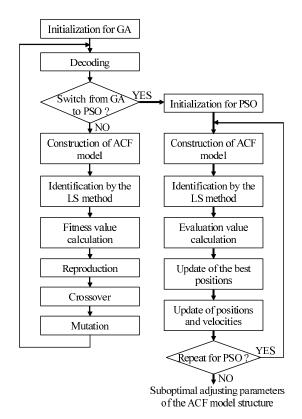


Fig. 5. Flow chart of the proposed hybrid method

V. NUMERICAL SIMULATIONS

Consider a system described by

$$\begin{cases} A(q^{-1})y(k) = B(q^{-1})x(k-1) + A(q^{-1})e(k) \\ x(k) = f(u(k)) \end{cases}$$

$$= \begin{cases} -2.0 & (-3.0 \le u(k) < -1.8) \\ u(k)/0.6 + 1.0 & (-1.8 \le u(k) < -0.6) \\ 0.0 & (-0.6 \le u(k) < 0.6) \\ u(k)/0.6 - 1.0 & (0.6 \le u(k) < 1.8) \\ 2.0 & (1.8 \le u(k) \le 3.0) \end{cases}$$

$$A(q^{-1}) = 1 + 0.8q^{-1} + 0.6q^{-2}$$

$$B(q^{-1}) = 0.4 + 0.2q^{-1}$$

This system has saturation and dead-zone nonlinearity. e(k)is white Gaussian noise with distribution N(0, 0.0015), where the noise-to-signal ratio is 5%. The number of input and output data is N = 300. The design parameters for the GA and PSO are chosen as follows:

GA parameters:

- 1) population size: Q = 50
- 2) search range of $\{\alpha_i\}$: $[\alpha_{min}, \alpha_{max}] = [-3.0, 3.0]$
- 3) search range of H: $[h_{min}, h_{max}] = [1.0, 200.0]$
- 4) crossover probability: $P_c = 0.8$
- 5) mutation probability: $P_m = 0.03$
- 6) maximum generation number: $g_{max} = 80$

PSO parameters:

1) particle size: Q = 50

TABLE I
SYSTEM PARAMETERS OF THE LINEAR DYNAMIC PART

	true values	estimates
a_1	0.8	0.779
a_2	0.6	0.593
b_1	0.2	0.188

TABLE II EVALUATION VALUES (AIC)

	Proposed hybrid	GA-based	PSO-based
best	-1.900e+3	-1.861e+3	-1.886e+3
mean	-1.852e+3	-1.829e+3	-1.848e+3
worst	-1.831e+3	-1.794e+3	-1.752e+3

Proposed hybrid	GA-based	PSO-based
0.0209	0.0223	0.0220

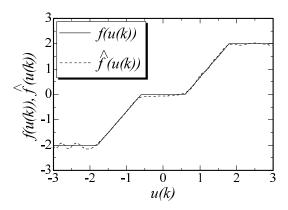


Fig. 6. True and estimated nonlinear functions

- 2) inertia factor: w = 0.7
- 3) acceleration coefficients: $c_1 = c_2 = 0.8$
- 4) maximum iteration number: $l_{max} = 120$

Estimates of the system parameters of the linear dynamic part are shown in Table I. In this table, the estimate of b_0 is omitted because the final estimated model is normalized by \hat{b}_0 . Clearly estimates by the proposed hybrid method is very close to the true parameters. The estimated nonlinear static function $\hat{f}(u(k))$ is shown in Fig. 6. We can confirm that $\hat{f}(u(k))$ matches the true nonlinear function f(u(k)) well on the given data region. Figure 7 shows the true output y(k), the output of the estimated model $\hat{y}(k)$ and $|y(k) - \hat{y}(k)|$, where the outputs were generated by the test inputs. It is clear that $\hat{y}(k)$ is very close to y(k).

For comparison, identification experiments by two other methods are carried out. One is the GA-based method in which the ACF model structure is determined by using only GA, and the other is the PSO-based method in which the ACF model structure is determined by using only PSO. Monte-

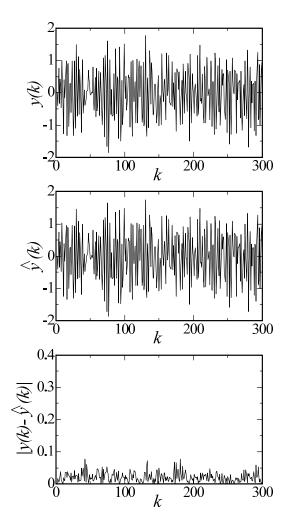


Fig. 7. True output, output by the estimated model, and difference between them

Carlo simulations of 20 experiments are implemented for the proposed hybrid method, the GA-based method and the PSO-based method. Table II shows the best, mean and worst evaluation values (AIC) of the 20 experiments for the three methods. The means of the output errors $\sum_{k=N_s+1}^{N_s+N} |y(k)-\hat{y}(k)|/N$ for the 20 experiments using the test data are shown in Table III. The performance of the proposed hybrid method is superior to those of other two methods.

VI. CONCLUSIONS

In this paper a method of model selection and identification of discrete-time Hammerstein systems by hybridization of the GA and PSO has been proposed. The nonlinear static part is approximately represented by the ACF model. Then the weighting parameters of the ACF and the system parameters of the linear dynamic part are estimated by the linear LS method. The adjusting parameters of the ACF model structure, i.e., the number and widths of the subdomains and the

shape of the ACF, are appropriately determined by the hybrid algorithm of the GA and PSO. Simulation results show that the identification by this method is easy in computation and can give accurate estimated models even in the presence of the measurement noises. Moreover the performance of the proposed hybrid algorithm is superior to those of the GA-based method and PSO-based method.

REFERENCES

- T. Liu, S. Boumaiza and F. M. Ghannouchi, Augmented Hammerstein predistorter for linearization of broad-band wireless transmitters, *IEEE Trans. Microwave Theory and Techniques*, Vol. 54, No.4, pp. 1340–1349, 2006.
- [2] F. Alonge, F. D'Ippolito, F. M. Raimondi and S. Tumminaro, Nonlinear modeling of dc/dc converters using the Hammerstein's approach, *IEEE Trans. Power Electronics*, Vol. 22, No. 4, pp. 1210–1221, 2007.
- [3] O. Nelles, Nonlinear System Identification, Springer, 2000.
- [4] S. A. Billings and S. Y. Fakhouri, Identification of systems containing linear dynamic and static nonlinear elements, *Automatica*, Vol. 18, No. 1, pp. 15–26, 1982.
- [5] H. Al-Duwaish and M. N. Karim, A new method for the identification of Hammerstein model, *Automatica*, Vol. 33, No. 10, pp. 1871–1875, 1997.
- [6] F. C. Kung and D. H. Shih, Analysis and identification of Hammerstein model non-linear delay systems using block-pulse function expansions, *Int. J. Control*, Vol. 43, No. 1, pp. 139–147, 1986.
- [7] S. Adachi and H. Murakami, Generalized predictive control system design based on non-linear identification by using Hammerstein model (in Japanese), *Trans. ISCIE*, Vol. 8, No. 3, pp. 115–121, 1995.
- [8] T. Hatanaka, K. Uosaki and M. Koga, Evolutionary computation approach to Hammerstein model identification, *Proc. 4th Asian Control Conf.*, pp. 1730–1735, 2002.
- [9] T. Hachino and H. Takata, Structure selection and identification of Hammerstein type nonlinear systems using automatic choosing function model and genetic algorithm, *IEICE Trans. Fundamentals of Electronics,* Communications and Computer Sciences, Vol. E88-A, No. 10, pp.2541– 2547, 2005.
- [10] D. E. Goldberg, Genetic Algorithms in Search, Optimization, and Machine Learning, Addison-Wesley, 1989.
- [11] J. Kennedy and R. C. Eberhart, Particle swarm optimization, *Proc. IEEE Int. Conf. Neural Networks*, pp. 1942–1948, 1995.
- [12] Y. Shi and R. C. Eberhart, Empirical study of particle swarm optimization, Proc. 1999 Congress on Evolutionary Computation, pp. 1945–1950, 1999
- [13] H. Yoshida, K. Kawata, Y. Fukuyama, S. Takayama and Y. Nakanishi, A particle swarm optimization for reactive power and voltage control considering voltage security assessment, *IEEE Trans. Power Syst.*, Vol. 15, No. 4, pp. 1232–1239, 2000.
- [14] A. Ide and K. Yasuda, A basic study of the adaptive particle swarm optimization (in Japanese), *IEEJ Trans. EIS*, Vol. 124, No. 2, pp. 550– 557, 2004.
- [15] M. Clerc and J. Kennedy, The particle swarm explosion, stability, and convergence in a multidimensional complex space, *IEEE Trans. Evolutionary Computation*, Vol. 6, No. 1, pp. 58–73, 2002.
- [16] V. Kadirkamanathan, K. Selvarajah and P. J. Fleming, Stability analysis of the particle dynamics in particle swarm optimizer, *IEEE Trans. Evolutionary Computation*, Vol. 10, No. 3, pp. 245–255, 2006.
- [17] H. Akaike, A new look at the statistical model identification, *IEEE Trans. Automatic Control*, Vol. 19, No. 6, pp. 716–723, 1974.
- [18] H. Takata, An automatic choosing control for nonlinear systems, Proc. of the 35th IEEE CDC, pp.3453–3458, 1996.

Tomohiro Hachino received the B.S., M.S. and Dr. Eng. degrees in electrical engineering from Kyushu Institute of Technology, in 1991, 1993 and 1996, respectively. He is currently an Associate Professor in the Department of Electrical and Electronics Engineering, Kagoshima University. His research interests include control and identification for nonlinear systems. Dr. Hachino is a member of IEEJ, SICE, and ISCIE.

Kenji Shimoda received the B.S. and M.S. degrees in electrical and electronics engineering from Kagoshima University in 2007 and 2009, respectively. In 2009, he joined Mitsubishi Electric Corporation. His research interests include nonlinear system identification and evolutionary computation.

Hitoshi Takata received the B.S. degree in electrical engineering from Kyushu Institute of Technology in 1968, and the M.S. and Dr. Eng. degrees in electrical engineering from Kyushu University, in 1970 and 1974, respectively. He is currently a Professor in the Department of Electrical and Electronics Engineering, Kagoshima University. His research interests include control, linearization, and identification for nonlinear systems. Dr. Takata is a member of IEEE, IEEJ, SICE, RISP, and ISCIE.