

# Comparison of different Channel Modeling Techniques used in the BPLC Systems

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**Abstract**—The paper compares different channel models used for modeling Broadband Power-Line Communication (BPLC) system. The models compared are Zimmermann and Dostert, Philipps, Anatory *et al* and Anatory *et al* generalized Transmission Line (TL) model. The validity of each model was compared in time domain with ATP-EMTP software which uses transmission line approach. It is found that for a power-line network with minimum number of branches all the models give similar signal/pulse time responses compared with ATP-EMTP software; however, Zimmermann and Dostert model indicates the same amplitude but different time delay. It is observed that when the numbers of branches are increased only generalized TL theory approach results are comparable with ATP-EMTP results. Also the Multi-Carrier Spread Spectrum (MC-SS) system was applied to check the implication of such behavior on the modulation schemes. It is observed that using Philipps on the underground cable can predict the performance up to 25dB better than other channel models which can misread the actual performance of the system. Also modified Zimmermann and Dostert under multipath can predict a better performance of about 5dB better than the actual predicted by Generalized TL theory. It is therefore suggested for a realistic BPLC system design and analyses the model based on generalized TL theory be used.

**Keywords**—Broadband Power line Channel Models, load impedance, Branched network.

## I. INTRODUCTION

FOR an efficient communication using the power line networks the channel's performance has to be evaluated to a greater accuracy. For this an appropriate and more accurate model is needed which can be used as a tool in the design of suitable communication equipments of PLC systems [1]. Researchers, in the recent past have attempted to come up with appropriate channel models, some of them being somewhat channel dependant, e.g. the indoor applications models by Banwell and Galli [2] and the low and medium voltage applications models by Zimmermann and Dostert [3], Hensen and Schulz [4], Philipps [5], etc. Some of the low and medium voltage channel models were extended to the indoor applications too. Some indoor models have been derived from measurements, e.g., Canete [6], [7]. Generally, the modeling adopted can be categorized as either time-domain models or frequency domain models [8]. It is however to be noted that the channels performance can be more accurately assessed

from channel transfer functions i.e., through frequency domain models. Another channel model based on transmission and reflection factors in conjunction with the propagation constants was proposed by Anatory *et al.* [9] for a PLC network with two conductor transmission line (TL) systems (one phase conductor and one return conductor).

The recent model by Anatory *et al.*, is based on the generalized TL theory approach [1] for determining the channel responses. It uses all the known principles of TL theory and the model is derived based on the modal analyses. Although, in the literature various types of channel models exist we feel that a comparative and sensitivity analyses is needed for more accurate channel performance studies within PLC. This paper compares Zimmermann and Dostert [3], Philipps [5], Anatory *et al.* [9] and Anatory *et al.* generalized TL approach [1]. The paper has considers cases whereby a power line channel with adjacent conductor return for different cases of distributed branches within the link between transmitting and receiving ends. The results in time domain are compared using widely used ATP-EMTP software by power engineers [10] which also uses transmission line approach.

## II. POWER-LINE CHANNEL MODELS

The transfer function of power line channel model proposed by Philipps [5] is given by (1). In (1)  $N$  is the number of possible signals flow paths, each path delayed by time  $\tau_i$  is multiplied by a complex factor  $\rho_i$ . The parameter  $\rho_i$  is the product of transmission and reflection factors. In (1) the parameters  $f$ ,  $c_o$ ,  $d_i$  and  $\epsilon_r$  are frequency, velocity of light, path length and relative permittivity respectively.

$$H(f) = \sum_{i=1}^N \rho_i e^{-j2\pi f \tau_i} \quad (1a)$$

$$\rho_i = |\rho_i| \cdot e^{j\phi_i}, \quad \phi_i = \arctan\left(\frac{\text{Im}(\rho_i)}{\text{Re}(\rho_i)}\right) \quad (1b)$$

$$\tau_i = \frac{d_i}{v_p} \quad (1c)$$

$$v_p = \frac{c_o}{\sqrt{\epsilon_r}} \quad (1d)$$

Zimmermann and Dostert [3] developed a channel model to account for the attenuation of the signal flow as given in (2). In (2) each path is characterized by weighting factor  $g_i$  which is the product of transmission and reflections factor

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with path length  $d_i$ . The attenuation factor is modeled by the parameters  $a_0$ ,  $a_1$  and  $k$ , which are obtained data from measurements also it employs top down approach. The model was extended to bottom-top model by deriving parameters from actual networks taking into consideration the connected loads [3]. The expressions are as shown in (3), where

$$t_{d_i} = \frac{d_i}{v_p} \text{ and } v_p = \frac{c_o}{\sqrt{\epsilon_r}} \text{ also } v_p = \frac{1}{\sqrt{L_e \cdot C_e}}. \text{ In 3b the}$$

parameters  $L_e$ ,  $C_e$ ,  $R$  and  $G$  are per unit length inductance, capacitance, resistance and conductance of a conductor respectively.  $\Gamma_{AD}$ ,  $\Gamma_{2C}$  and  $\rho_{2D}$  are transmission factor from point A towards D, transmission factor along line 2 to point C, reflection factor along line 2 to point D respectively.

$$H(f) = \sum_{i=1}^N g_i e^{-(a_0 + a_1 f^k) \cdot d_i} e^{-j2\pi f \frac{d_i}{v_p}} \quad (2)$$

$$H(f) = \sum_{i=1}^N g_i''(f) A''(f_i, d) \cdot \exp(-j2\pi f \tau_{d_i}) \quad (3a)$$

$$A''(f_i, d_i) = \exp(-\sqrt{(R + j\omega L_e)(G + j\omega C_e)}) \cdot d_i \quad (3b)$$

$$g_i'' = \Gamma_{AD} \rho_{2D}^{(i-1)} \Gamma_{2C}^{(i-1)} \quad (3c)$$

Anatory *et al.* [9] developed a generalized case applicable to any line configuration with distributed branches along the line from transmission point to the receiver. The transfer function of such network is given by (4a). In (4a) the parameters  $M_T$  is the total number of distributed nodes,  $d$  is any referenced node ( $1 \dots M_T$ ),  $H_{mnd}(f)$  is the transfer function between line  $n$  to a referenced load  $m$  at a referenced node  $d$ . For more information about such parameters the reader are advised to use the papers such as [9].

$$H_{mM_T}(f) = \prod_{d=1}^{M_T} \sum_{M=1}^{L_n} \sum_{n=1}^{N_T} T_{Lnd} \alpha_{mnd} H_{mnd}(f) \quad n \neq m \quad (4a)$$

$$\alpha_{mnd} = P_{Lnd}^{M-1} \rho_{nm}^{M-1} e^{-\gamma_{nm} (2(M-1)\ell_{nd})} \quad (4b)$$

$$P_{Lnd} = \begin{cases} \rho_s & d = n = 1(\text{source}) \\ \rho_{Lnd}, & \text{otherwise} \end{cases} \quad (4c)$$

The recent generalized transmission line (TL) theory was developed by Anatory *et al.* [1]. Note that the procedure for the obtaining the transfer function is the same i.e. writing the voltage and current boundary conditions at all the nodes and solving for the unknown modal currents. The transfer function for the voltage between any load point  $Z_{nm}$  and the sending end is given by (5a). In (5a)  $Z_{Cnm}$ ,  $Z_{nm}$ ,  $\gamma_{nm}$ ,  $L_{nm}$  and  $L_n$  are characteristic impedance of line segment  $nm$ , terminal load impedance of line  $nm$ , propagation constant of line segment  $nm$ , shortest length of line segment  $nm$ , and shortest line length from the sending end to the node  $n$  under

consideration, respectively. Note that all parameters with  $nm$  means the consideration is at the node.

$$H_{nm}(f) = \frac{Z_{C11} + Z_s}{Z_{C11}} Z_{Cnm} (e^{-\gamma_{nm} L_{nm}} \beta_{nm} + e^{\gamma_{nm} L_{nm}}) A_{nm} \frac{1}{A_1^+ \beta_1 + A_1^-} \quad (5a)$$

$$\beta_{nm} = \frac{C_{nm}^- - P_n B_{nm}^-}{C_{nm}^+ + P_n B_{nm}^+} \quad (5b)$$

$$P_n = \frac{e_{nm(1)}}{a_{nm(1)}} + \frac{e_{nm(2)}}{a_{nm(2)}} \quad (5c)$$

$$\beta_{nm} = \frac{(1 - Z_{Cnm}/Z_{nm}) e^{-\gamma_{nm} L_{nm}}}{(1 + Z_{Cnm}/Z_{nm}) e^{\gamma_{nm} L_{nm}}} \quad (5d)$$

$$A_{nm} = \frac{a_n a_{n-1} \dots a_1}{a_{nm} a_{(n-1)(n)} a_{(n-2)(n-1)} \dots a_{12}} \quad (5e)$$

$$a_{nm} = B_{nm}^+ \beta_n + B_{nm}^- \quad (5f)$$

$$e_{nm} = C_{nm}^- - C_{nm}^+ \beta_n \quad (5g)$$

$$\beta_n = \begin{cases} \beta_{(n+1)(n+1)} & \text{to node } n+1 \\ \beta_{nm} & \text{to load } m \end{cases} \quad (5h)$$

$$a_{mn} = B_{nm}^+ \beta_{nm} + B_{nm}^- \quad (5i)$$

$$a_n = B_{nm}^+ \beta_{nm} + B_{nm}^- \quad (5j)$$

$$B_{nm}^+ = Z_{Cnm} e^{-\gamma_{nm} L_n} \quad (5k)$$

$$B_{nm}^- = Z_{Cnm} e^{\gamma_{nm} L_n} \quad (5l)$$

$$B_{nm}^+ = Z_{Cnm} e^{-\gamma_{nm} L_n} \quad (5m)$$

$$B_{nm}^- = Z_{Cnm} e^{\gamma_{nm} L_n} \quad (5n)$$

$$C_{nm}^+ = e^{-\gamma_{nm} L_n} \quad (5o)$$

$$C_{nm}^- = e^{\gamma_{nm} L_n} \quad (5p)$$

$$B_{nm}^+ = Z_{Cnm} e^{-\gamma_{nm} L_n} \quad (5q)$$

$$B_{nm}^- = Z_{Cnm} e^{\gamma_{nm} L_n} \quad (5r)$$

$$C_{nm}^+ = e^{-\gamma_{nm} L_n} \quad (5s)$$

$$C_{nm}^- = e^{\gamma_{nm} L_n} \quad (5t)$$

In equation (5a),  $A_1^+ = Z_{C11} + Z_s$ , and  $A_1^- = Z_{C11} - Z_s$ .

For all models the received signals  $V_R$  can be obtained using (6), where by  $Z_L$ ,  $H(f)$  and  $Z_s$  are the load impedance at the receiving terminal, transfer function and load impedance of the transmitter respectively.

$$V_R(f) = H(f) * \left( \frac{Z_L}{Z_L + Z_s} \right) V_s \quad (6)$$

### III. POWER LINE NETWORK WITH ADJACENT RETURN CONDUCTOR- LOSSLESS CASES

#### A. Power line Network Configuration with one branch

Consider power line network as shown in Fig. 1,  $Z_S$ ,  $V_S$ ,  $Z_{L1}$  and  $Z_{L2}$  are source impedance, source voltage, load impedance at node C and load impedance at load D, respectively. The length of line segment AB, BD and BC was considered as 60m, 200m and 100m respectively. Per unit length inductances and capacitances was taken as  $0.44388\mu\text{H/m}$  and  $0.61734\text{pF/m}$  respectively for all the line segments. A 2V rectangular pulse with width of  $1\mu\text{s}$  shifted by  $0.5\mu\text{s}$  was considered as the voltage source injection and was applied to the Phillips [3], Zimmermann and Dostert [5], Anatory *et al.* [9] and Anatory *et al.* generalizes TL theory [1] models. In the cases of Phillips, Zimmermann and Dostert models 10 paths was considered. In Anatory *et al.* [9] model 10 total numbers of reflections were considered. In the simulations  $Z_{L1}$  was kept open while  $Z_S$  and  $Z_{L2}$  were terminated in the characteristic impedance. The voltage was calculated across  $Z_{L2}$ .

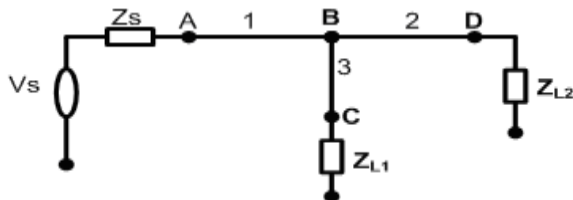


Fig. 1 Power line Network Configuration with one branch

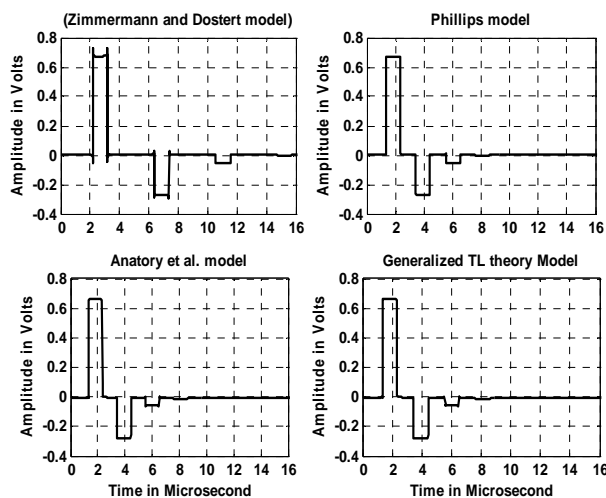


Fig. 2a Comparisons for power line channel models for a power line network with one branch.

Fig. 2a shows the simulation results for Phillips, Zimmermann and Dostert, Anatory *et al.* and the Anatory *et al.* generalized TL theory models. It is observed all the models above have similar results while Zimmermann and Dostert [5] model shows deviation along the time axis, i.e., the instances of voltage peaks or dips (time delays of signals) are not consistent with other models. The validity of the models for this configuration was implemented in ATP-EMTP software and the corresponding results are shown in Fig. 2b, which confirms that Phillips, Anatory *et al.* and generalized TL theory models are consistent. Then let us consider a case of

power line configuration with two distributed branches in the link between transmitting and receiving ends.

#### B. Power line Network Configuration with two Distributed branches

Consider power line network as shown in Fig. 3, wherein  $Z_S$ ,  $V_S$ ,  $Z_{L1}$ ,  $Z_{L2}$  and  $Z_{L3}$  are source impedance, source voltage, load impedance at node E, load impedance at node D and load impedance at load F respectively. The length of line segment AB, BC, BE, CD and CF are 200m. Per unit length inductances and capacitances of all line segments are the same as in previous case. The same rectangular pulse as in previous case was used as voltage source in the Phillips [3], Zimmermann and Dostert [5], Anatory *et al.* [9] and Anatory *et al.* generalized TL theory model [20]. In the cases of Phillips, Zimmermann and Dostert models 10 paths were considered. In Anatory *et al.* [9] model 10 total numbers of reflections were considered. In the investigation,  $Z_{L1}$  and  $Z_{L2}$  were kept at  $20\Omega$  and  $50\text{k}\Omega$ , respectively; while  $Z_S$  and  $Z_{L3}$  were terminated in  $85\Omega$ . Fig. 4a shows the simulations results for all cases. It is clearly observed that the Phillips and Anatory *et al.* models have similar results while Anatory *et al.* generalized TL theory model predicts different responses after about  $10\mu\text{s}$ .

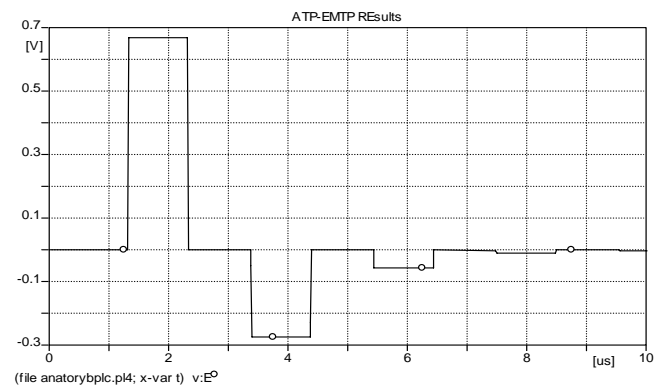


Fig. 2b Simulations using ATP-EMTP software for a power line network with one branch

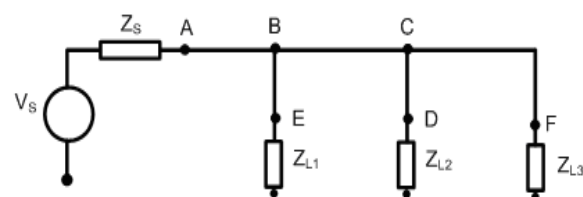


Fig. 3 Network Configuration with distributed branches

Zimmermann and Dostert model indicates similar amplitude but with different time delays as observed in the previous case. The configuration was implemented in ATP-EMTP software, and the corresponding simulations are shown in Fig. 4b. It is evident that the generalized TL theory model predictions are consistent with the ATP-EMTP result (compare the amplitudes of all models even after  $10\mu\text{s}$ ), which indicates that the generalized TL theory model is more accurate, compared to other models. From the above observations, it is found that the Zimmermann and Dostert

model might be incapable of predicting the delay spread aspects in BPLC based on TL theory.

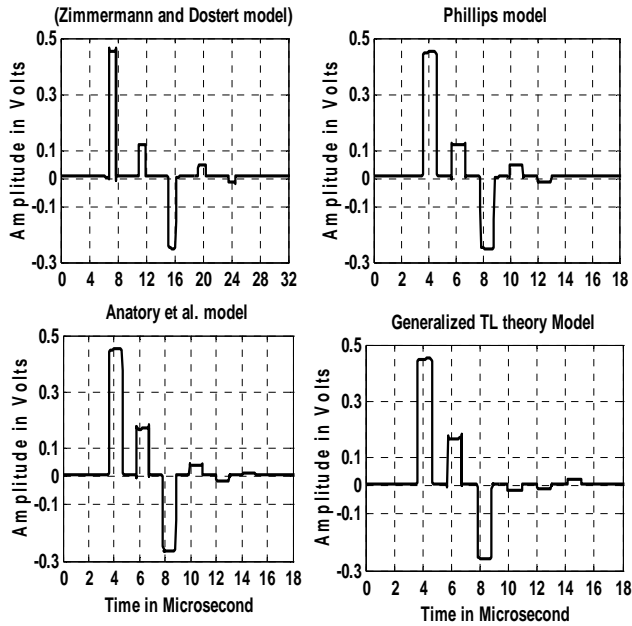


Fig 4a Comparisons for power line channel models for a power line network with two distributed branches

#### IV. POWER LINE NETWORK WITH ADJACENT RETURN CONDUCTOR- LOSS CASES

Consider power line network cables with configuration as shown in fig. 3, wherein  $Z_s$ ,  $V_s$ ,  $Z_{L1}$ ,  $Z_{L2}$  and  $Z_{L3}$  are source impedance, source voltage, load impedance at node E, load impedance at node D and load impedance at load F respectively. The length of line segment AB, BC, BE, CD and CF are 200m. Per unit length inductances and capacitances of all line segments are taken as  $0.44388\mu\text{H/m}$  and  $0.61734\text{pF/m}$  respectively. The conductance and resistance expression was considered as (7) and (8) respectively. In (7) and (8) the parameters  $\delta$ ,  $\sigma$ ,  $\mu$  and  $r$  are the depth factor, conductivity of the conductor, permeability of the conductor and radius of the conductor respectively. In this case the conductivity and conductance was considered as  $5.8e7$  and  $4e-7$  respectively. the radius of the conductor connecting the sending and receiving ends were considered as  $0.69099\text{mm}$ , while for branched cables was considered as  $0.39894\text{mm}$ . In the investigation  $Z_{L1}$  and  $Z_{L2}$  were kept at  $10\text{M}\Omega$  respectively, while  $Z_s$  and  $Z_{L3}$  were terminated in  $85\Omega$ . A  $2\text{V}$  rectangular pulse with width of  $1\mu\text{s}$  shifted by  $0.5\mu\text{s}$  voltage source was injected in the Phillips [3], Zimmermann and Dostert [3], Anatory *et al* [9] and generalize TL theory model [1]. In the cases of Phillips, Zimmermann and Dostert models 10 paths were considered. In Anatory *et al* [9] model 10 total numbers of reflections were considered. Fig. 5 shows the simulations results for all cases. It is observed that Phillips model is not affected by loss cases of a power line channel. Other models are affected by loss cases. This concludes that in Phillips models apart from giving incorrect amplitudes at the late time

the model is not also appropriate for lossy cases which is real case in power line network infrastructure.

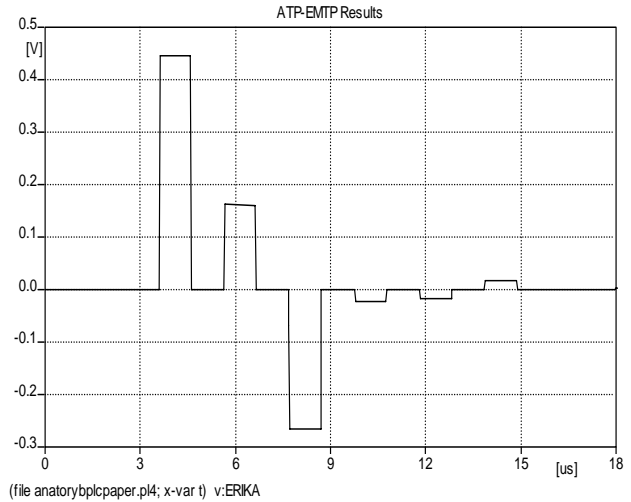


Fig. 4b Simulations using ATP-EMTP software for a power line network with two distributed branches

$$G = 2\pi f C \tan \delta \quad (7a)$$

$$\delta = \frac{1}{\sqrt{\pi f \mu \sigma}} \quad (7b)$$

$$R = \frac{1}{\pi \sigma r^2} + \sqrt{2j\pi f} \cdot \sqrt{\frac{\pi \mu / \sigma}{2\pi r}} \quad (8)$$

#### V. IMPROVED ZIMMERMANN AND DOSTERT MODEL FOR POWER LINE NETWORK WITH ADJACENT RETURN CONDUCTOR- LOSS CASES

In this section we looked at the way Zimmermann and Dostert [3] can be improved. Consider power line network cables with configuration as shown in fig. 3, wherein  $Z_s$ ,  $V_s$ ,  $Z_{L1}$ ,  $Z_{L2}$  and  $Z_{L3}$  are source impedance, source voltage, load impedance at node E, load impedance at node D and load impedance at load F respectively. The length of line segment AB, BC, BE, CD and CF are 200m. Per unit length inductances and capacitances of all line segments are taken as  $0.44388\mu\text{H/m}$  and  $0.61734\text{pF/m}$  respectively. The conductance and resistance expression was considered as (7) and (8) respectively. In (7) and (8) the parameters  $\delta$ ,  $\sigma$ ,  $\mu$  and  $r$  are the depth factor, conductivity of the conductor, permeability of the conductor and radius of the conductor respectively. In this case the conductivity and conductance was considered as  $5.8e7$  and  $4e-7$  respectively. The radius of the conductor connecting the sending and receiving ends was considered as  $0.69099\text{mm}$ , while for branched cables were considered as  $0.39894\text{mm}$ . In the investigation  $Z_{L1}$  and  $Z_{L2}$  were kept at  $10\text{M}\Omega$  respectively, while  $Z_s$  and  $Z_{L3}$  were terminated in  $85\Omega$ . Zimmermann and Dostert [3] model was implemented as in (7) and (8). Note that in (7) and (8) the path length in the delay parameter has

been removed. A 2V rectangular pulse with width of 1μs shifted by 0.5μs voltage source was injected in the Phillips [3], Zimmermann and Dostert [5], Anatory *et al* [9] and generalize TL theory model [1]. In the cases of Phillips, Zimmermann and Dostert models 10 paths were considered. In Anatory *et al* [9] model 10 total numbers of reflections were considered. Fig. 6 shows the simulations results for all cases. It is observed in terms of time delay all models are comparable which indicates that in Zimmermann and Dostert [3] model the delay parameters should be removed.

$$H_{\text{Improved}}(f) = \sum_{i=1}^N g_i e^{-(a_0 + a_1 f^k) \cdot d_i} e^{-j2\pi f \frac{1}{v_p}} \quad (9)$$

$$H_{\text{Improved}}(f) = \sum_{i=1}^N g_i(f) A''(f_i, d) \cdot \exp(-j2\pi f \frac{1}{v_p}) \quad (10a)$$

$$A''(f_i, d_i) = \exp(-\sqrt{(R + j\omega L_e)(G + j\omega C_e)}) \cdot d_i \quad (10b)$$

$$g''_i = \Gamma_{AD} \rho_{2D}^{(i-1)} \Gamma_{2C}^{(i-1)} \quad (10c)$$

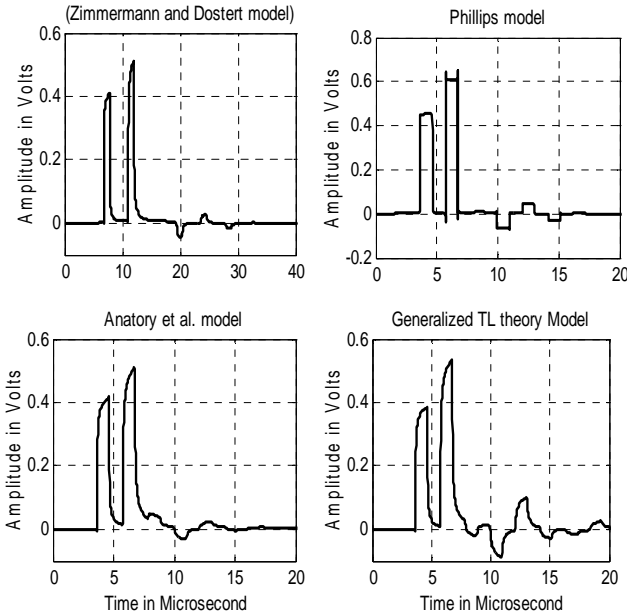


Fig. 5 Comparisons for power line channel models for a power line network with two distributed branches with terminal loads terminated in 10M-ohms [11]

## VI. MODEL COMPARISONS FOR UNDERGROUND CABLES POWERLINE LINK

Consider power line network cables with configuration as shown in fig. 7, wherein  $Z_s$ ,  $V_s$ ,  $Z_L$  are source impedance, source voltage, and load impedance at the receiving end respectively. The number of branches between A and J was eight and distributed equally between sending end and receiving ends. The lengths of branched line are 15m. Per unit length inductances and capacitances of all line segments between A and J are taken as 0.32735μH/m and 0.27191pF/m

respectively while the per unit length inductances and capacitances for branched line length are taken as 0.45179μH/m and 0.19702pF/m respectively. The conductance and resistance expression was considered as (7) and (8) respectively. The radius of the conductor connecting the sending and receiving ends was considered as 0.69099mm, while for branched cables were considered as 0.39894mm. In the investigation all terminal loads including  $Z_s$  and  $Z_L$  were terminated in characteristics impedances.

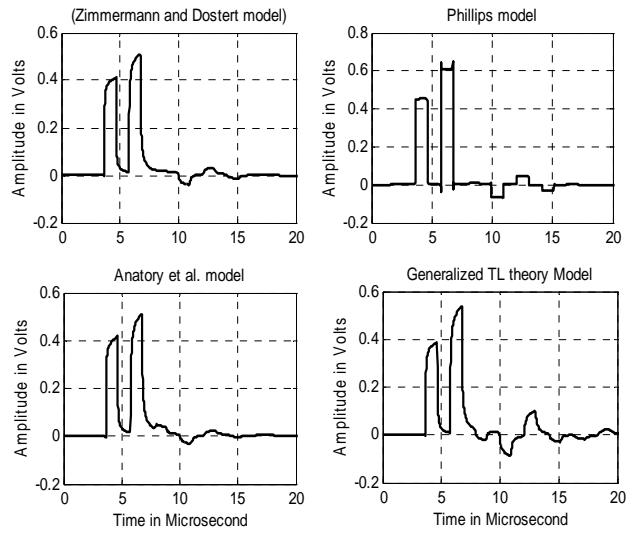


Fig. 6 Comparisons for power line channel models with Zimmermann and Dostert [5] improved for a power line network with two distributed branches with terminal loads terminated in 10M-ohms [11]

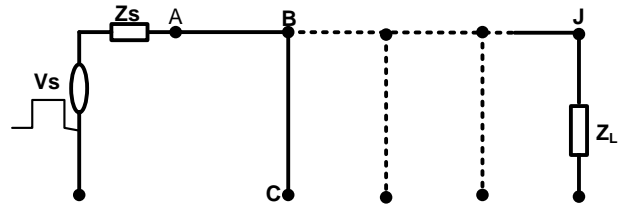


Fig. 7 Network Configuration with distributed branches

### A. Frequency responses

Firstly all terminal loads were terminated in the characteristics impedances then in higher impedances. The higher impedances were considered as 1KΩ. Phillips [5], Zimmermann and Dostert [3], Anatory *et al* [9] and generalize TL theory model [1] were applied in the links. In the cases of Phillips, Zimmermann and Dostert models 10 paths were considered. In Anatory *et al* [9] model 10 total numbers of reflections were considered. Figure 8 is the channel frequency responses for underground cables when all terminals are terminated in the characteristics impedances. It can be observed that Phillips model doesn't attenuate with frequencies while the rest of the models attenuate with frequencies. The position of notches and peaks for all models are the same. At higher frequencies, i.e at 20MHz there is difference of 20dB using generalized TL theory model in

comparison with modified Zimmermann and Dostert [3] and Anatory *et al* [9]. Figure 9 shows the frequency responses for all models when all branches for a configuration shown in figure 7 are terminated in 1K $\Omega$ . It can be observed that except Phillips Model which doesn't attenuate with frequency the rest models attenuate with frequencies. However it has been observed that at higher frequencies generalized TL theory model is attenuating more compared to other models. For example at 25MHz generalized TL theory model, modified Zimmermann and Dostert [3] and Anatory *et al* is at -55dB, -45dB and -50dB respectively.

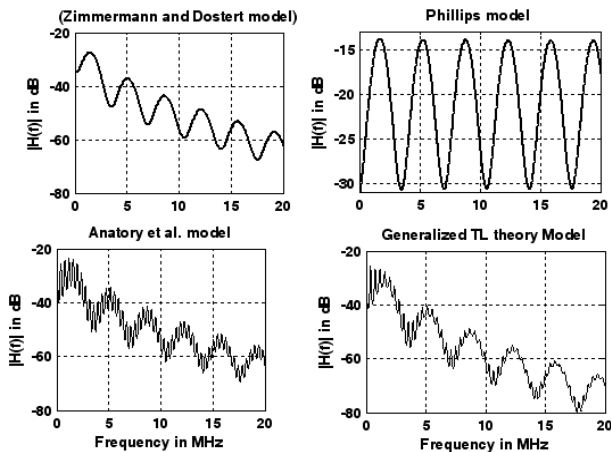


Fig. 8 Frequency Response Comparisons for underground power line channel models with improved Zimmermann and Dostert [3] for a power line network with eight distributed branches with terminal loads terminated in characteristics impedances [11]

## VII. CONCLUSION

In this paper different power line channel models namely, Phillips [3], Zimmermann and Dostert [5], Anatory *et al.* [7] and Anatory *et al.* generalized TL theory model [1,12] are compared. The comparison is done in time domain and ATP-EMTP software is used to check the validity of the model responses. Two important findings are:

1. It is found that Zimmermann and Dostert model predicts incorrect time delays for the responses even though the amplitudes are comparable with other models.
2. As a number of branches increase all the models deviate except the Anatory *et al.* generalized TL theory model [1] is comparable with that of the ATP-EMTP software predictions which is based of transmission line theory.
3. For underground cables Phillips model is inappropriate since it doesn't capture the attenuation factors of the cable which can lead to inappropriate performances using multicarrier systems.
4. In the case of branched power line channel Zimmermann and Dostert model is not suitable since it can predict inappropriate performances.

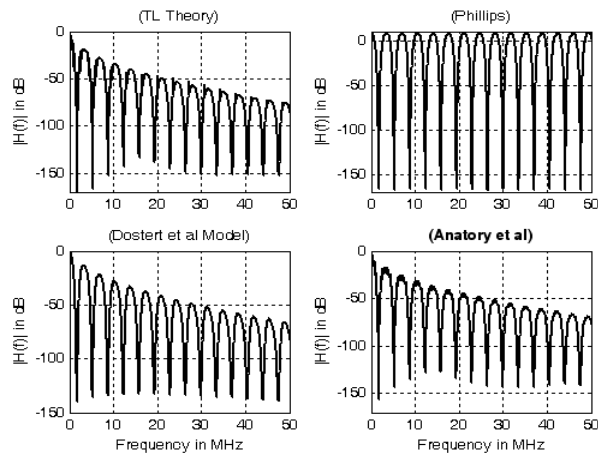


Fig. 9 Frequency Response Comparisons for power line channel models with improved Zimmermann and Dostert [3] for a power line network with eight distributed branches with terminal loads terminated in 1K-ohms [11]

## REFERENCES

- [1] Justinian Anatory, N. Theethayi, and R. Thottappillil "Power-Line Communication Channel Model for Interconnected Networks - Part I: Two Conductor System", *IEEE Transactions on Power Delivery*, Vol. 24, No. 1, January 2009.
- [2] T. Banwell and S. Galli, "A Novel Approach to the Modeling of the Indoor Powerline Channel -Part I: Circuit Analysis and Companion Model", *IEEE Trans. On Power Delivery*, vol. 20, no.2, April 2005. pp. 655-663.
- [3] M. Zimmermann and K. Dostert; "A Multipath Model for the Powerline Channel", *IEEE Trans. On Communications*, vol. 50, No. 4, April 2002. pp. 553-559
- [4] C. Hensen and W. Schulz, "Time dependence of the channel characteristics of low voltage power-lines and its effects on hardware implementation," *AEU Int. J. Electron. Commun.*, vol. 54, no. 1, pp. 23-32, Feb. 2000.
- [5] H. Philipps, "Modeling of power line communication channels," in *Proc. 3rd Int. Symp. Power-Line Communications Applications*, Lancaster, U.K., 1999, pp. 14-21.
- [6] F. J. Canete, L. Diez, J. A. Cortes, J. T. Entrambasaguas, "Modelling and Evaluation of the Indoor Powerline Transmission Medium", *IEEE Communications Magazine*, April 2003.
- [7] F. J. Canete, L. Diez, J. A. Cortes, J. T. Entrambasaguas, "Broadband Modeling of Indoor Power-Line Channels", *IEEE Trans. On Consumer Electronics*, Vol. 48, No. 1, Feb. 2002.
- [8] X. Ding and J. Meng, "Channel Estimation and Simulation of an Indoor Power-Line Network via a Recursive Time-Domain Solution", *IEEE Trans Power Del.*, vol. 24, no. 1, Jan. 2009.
- [9] J. Anatory, M.M. Kissaka and N.H. Mvungi, "Channel Model for Broadband Powerline Communication", *IEEE Trans. On Power Delivery*, Vol. 22, no. 1, January 2007, pp. 135-141.
- [10] H. W. Dommel, *Electromagnetic Transients Program (EMTP Theory Book)*. Portland, OR: Bonneville Power Administration, 1988
- [11] Justinian Anatory and Nelson Theethayi, *Broadband Power-Line Communication Systems: Theory and Applications*, WIT Press, UK, May, 2010, ISBN 978-1-84564-416-1
- [12] J. Anatory, N. Theethayi, and R. Thottappillil and N.H. Mvungi "A Broadband Power-Line Communication System Design Scheme for Typical Tanzanian Low Voltage Network", *IEEE Transactions on Power Delivery*, Vol. 24, No. 3, pp. 1218-1224, July, 2009

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